

### LM725

## **Operational Amplifier**

### **General Description**

The LM725/LM725A/LM725C are operational amplifiers featuring superior performance in applications where low noise, low drift, and accurate closed-loop gain are required. With high common mode rejection and offset null capability, it is especially suited for low level instrumentation applications over a wide supply voltage range.

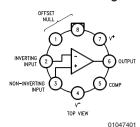
The LM725A has tightened electrical performance with higher input accuracy and like the LM725, is guaranteed over a -55°C to +125°C temperature range. The LM725C has slightly relaxed specifications and has its performance guaranteed over a 0°C to 70°C temperature range.

#### **Features**

- High open loop gain 3,000,000
- Low input voltage drift 0.6 µV/°C
- High common mode rejection 120 dB
- Low input noise current 0.15 pA/√Hz
- Low input offset current 2 nA
- High input voltage range ±14V
- Wide power supply range ±3V to ±22V
- Offset null capability
- Output short circuit protection

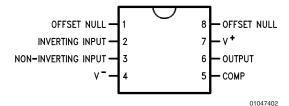
### **Connection Diagrams**

#### Metal Can Package



Order Number LM725H/883, LM725CH or LM725AH/883 See NS Package Number H08C

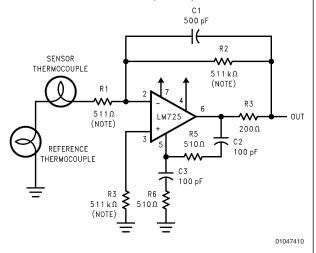
#### **Dual-In-Line Package**



Order Number LM725CN See NS Package Number N08E

### **Typical Applications**

#### Thermocouple Amplifier



## **Absolute Maximum Ratings** (Note 1)

If Military/Aerospace specified devices are required, please contact the National Semiconductor Sales Office/ Distributors for availability and specifications.

Supply Voltage  $\pm 22V$ Internal Power Dissipation (Note 2) 500 mW
Differential Input Voltage  $\pm 5V$ Input Voltage (Note 3)  $\pm 22V$ Storage Temperature Range  $-65^{\circ}$ C to  $+150^{\circ}$ C

Lead Temperature(Soldering, 10 Sec.) $260^{\circ}$ CMaximum Junction Temperature $150^{\circ}$ COperating Temperature Range $T_{A(MIN)}$  $T_{A(MAX)}$ LM725 $-55^{\circ}$ Cto  $+125^{\circ}$ CLM725A $-55^{\circ}$ Cto  $+125^{\circ}$ CLM725C $0^{\circ}$ Cto  $+70^{\circ}$ C

### **Electrical Characteristics** (Note 4)

		LM725A			LM725			LM725C			
Parameter	Conditions	Min	Тур	Max	Min	Тур	Max	Min	Тур	Max	Units
Input Offset Voltage	$T_A = 25^{\circ}C$ ,			0.5		0.5	1.0		0.5	2.5	mV
(Without External Trim)	$R_S \le 10 \text{ k}\Omega$										
Input Offset Current	$T_A = 25^{\circ}C$		2.0	5.0		2.0	20		2.0	35	nA
Input Bias Current	$T_A = 25^{\circ}C$		42	80		42	100		42	125	nA
Input Noise Voltage	$T_A = 25^{\circ}C$										
	f <sub>o</sub> = 10 Hz		15			15			15		nV/√ <del>Hz</del>
	f <sub>o</sub> = 100 Hz		9.0			9.0			9.0		nV/√ <del>Hz</del>
	$f_o = 1 \text{ kHz}$		8.0			8.0			8.0		nV/√ <del>Hz</del>
Input Noise Current	T <sub>A</sub> = 25°C										
	f <sub>o</sub> = 10 Hz		1.0			1.0			1.0		pA/√ <del>Hz</del>
	f <sub>o</sub> = 100 Hz		0.3			0.3			0.3		pA/√ <del>Hz</del>
	$f_o = 1 \text{ kHz}$		0.15			0.15			0.15		pA/√ <del>Hz</del>
Input Resistance	T <sub>A</sub> = 25°C		1.5			1.5			1.5		MΩ
Input Voltage Range	T <sub>A</sub> = 25°C	±13.5	±14		±13.5	±14		±13.5	±14		V
Large Signal Voltage Gain	$T_A = 25^{\circ}C,$										
	$R_L \ge 2 k\Omega$ ,	1000	3000		1000	3000		250	3000		V/mV
	$V_{OUT} = \pm 10V$										
Common-Mode	$T_A = 25^{\circ}C,$	120			110	120		94	120		dB
Rejection Ratio	$R_S \le 10 \text{ k}\Omega$										
Power Supply	T <sub>A</sub> = 25°C,		2.0	5.0		2.0	10		2.0	35	μV/V
Rejection Ratio	$R_S \le 10 \text{ k}\Omega$										
Output Voltage Swing	T <sub>A</sub> = 25°C,										
	$R_L \ge 10 \text{ k}\Omega$	±12.5	±13.5		±12	±13.5		±12	±13.5		V
	$R_L \ge 2 k\Omega$	±12.0	±13.5		±10	±13.5		±10	±13.5		V
Power Consumption	T <sub>A</sub> = 25°C		80	105		80	105		80	150	mW
Input Offset Voltage	$R_S \le 10 \text{ k}\Omega$			0.7			1.5			3.5	mV
(Without External Trim)											
Average Input Offset	$R_S = 50\Omega$										
Voltage Drift				2.0		2.0	5.0		2.0		μV/°C
(Without External Trim)											
Average Input Offset	$R_S = 50\Omega$										
Voltage Drift			0.6	1.0		0.6			0.6		μV/°C
(With External Trim)											-
Input Offset Current	$T_A = T_{MAX}$		1.2	4.0		1.2	20		1.2	35	nA
•	$T_A = T_{MIN}$		7.5	18.0		7.5	40		4.0	50	nA
Average Input Offset	, , , , , , , , , , , , , , , , , , ,	1	35	90		35	150		10		pA/°C
Current Drift											
Input Bias Current	$T_A = T_{MAX}$		20	70		20	100			125	nA
•	$T_A = T_{MIN}$		80	180		80	200			250	nA
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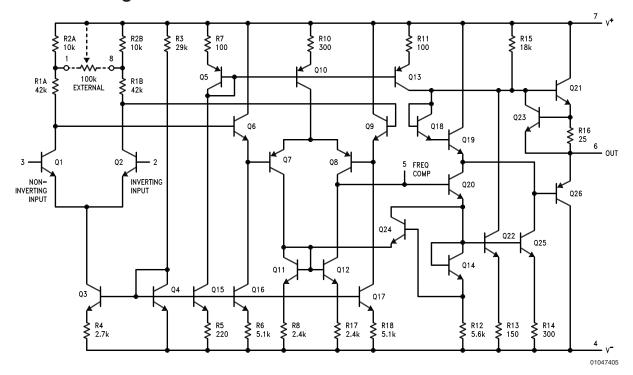
### Electrical Characteristics (Note 4) (Continued)

		LM725A		LM725		LM725C		
Parameter	Conditions	Min	Тур Мах	Min	Тур Мах	Min	Тур Ма	Units
Large Signal Voltage Gain	$R_L \ge 2 k\Omega$							
	$T_A = T_{MAX}$	1,000,000		1,000,000		125,000		V/V
	$R_L \ge 2 k\Omega$							
	$T_A = T_{MIN}$	500,000		250,000		125,000		V/V
Common-Mode	$R_S \le 10 \text{ k}\Omega$	110		100			115	dB
Rejection Ratio								
Power Supply	$R_S \le 10 \text{ k}\Omega$		8.0		20		20	μV/V
Rejection Ratio								
Output Voltage Swing	$R_L \ge 2 \ k\Omega$	±12		±10		±10		V

Note 1: "Absolute Maximum Ratings" indicate limits beyond which damage to the device may occur. Operating Ratings indicate conditions for which the device is functional, but do not guarantee specific performance limits.

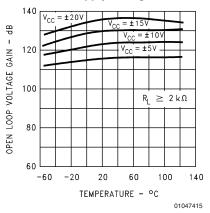
- Note 2: Derate at 150°C/W for operation at ambient temperatures above 75°C.
- $\textbf{Note 3:} \ \ \text{For supply voltages less than $\pm 22$V}, \ \text{the absolute maximum input voltage is equal to the supply voltage}.$
- Note 4: These specifications apply for  $V_S = \pm 15V$  unless otherwise specified.
- Note 5: For Military electrical specifications RETS725AX are available for LM725AH and RETS725X are available for LM725H.

### **Schematic Diagram**

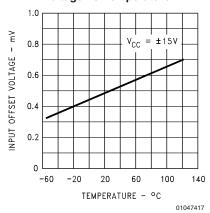


## **Typical Performance Characteristics**

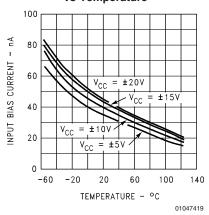
#### Voltage Gain vs Temperature for Supply Voltages



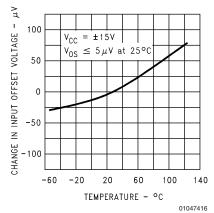
## Untrimmed Input Offset Voltage vs Temperature



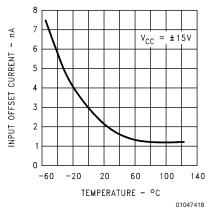
## Input Bias Current vs Temperature



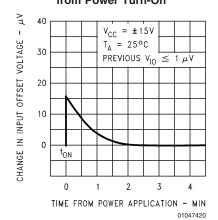
## Change in Trimmed Input Offset Voltage vs Temperature



## Input Offset Current vs Temperature

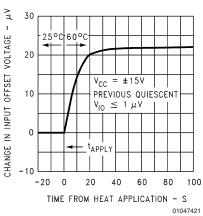


# Stabilization Time of Input Offset Voltage from Power Turn-On

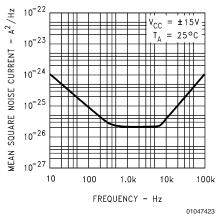


### **Typical Performance Characteristics** (Continued)

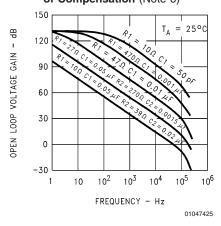
#### Change in Input Offset Voltage Due to Thermal Shock vs Time



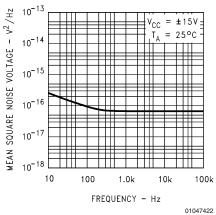
#### Input Noise Current vs Frequency



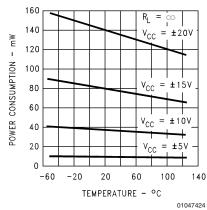
#### Open Loop Frequency Response for Values of Compensation (Note 6)



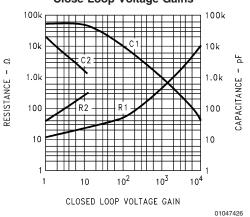
## Input Noise Voltage vs Frequency



## Power Consumption vs Temperature



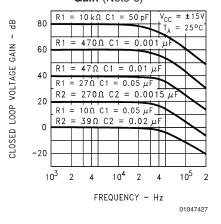
#### Values for Suggested Compensation Networks vs Various Close Loop Voltage Gains



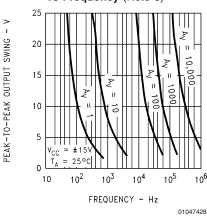
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## **Typical Performance Characteristics** (Continued)

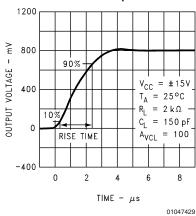
#### Frequency Response for Various Close Loop Gain (Note 6)



## Output Voltage Swing vs Frequency (Note 6)

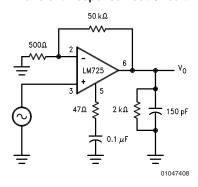


#### **Transient Response**



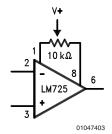
**Note 6:** Performance is shown using recommended compensation networks.

#### **Transient Response Test Circuit**

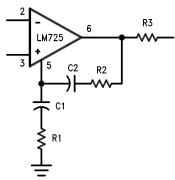


## **Auxiliary Circuits**

### Voltage Offset Null Circuit



# Frequency Compensation Circuit



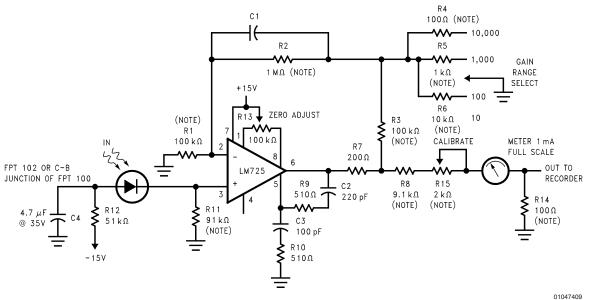
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#### **Compensation Component Values**

A <sub>V</sub>	R <sub>1</sub>	C <sub>1</sub>	R <sub>2</sub>	$C_2$
	<b>(</b> Ω <b>)</b>	(μ <b>F</b> )	<b>(</b> Ω <b>)</b>	(μ <b>F</b> )
10,000	10k	50 pF		
1,000	470	0.001		
100	47	0.01		
10	27	0.05	270	0.0015
1	10	0.05	39	0.02

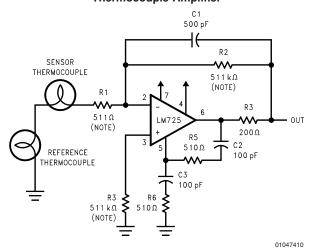
## **Typical Applications**

#### Photodiode Amplifier



DC Gains = 10,000; 1,000; 100; and 10
Bandwidth = Determined by value of C1

#### Thermocouple Amplifier

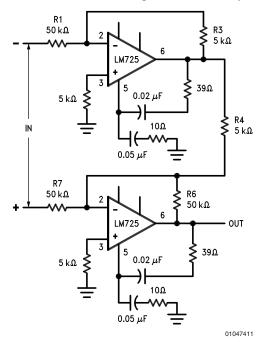


$$\begin{array}{l} \frac{R2}{R5} = \frac{R6}{R7} \text{ for best CMR} \\ \\ R1 = R4 \\ R2 = R5 \\ \\ Gain = \frac{R6}{R2} + \left(\frac{2R1}{R3}\right) \\ \\ DC \text{ Gain} = 1000 \\ \\ Bandwidth = DC \text{ to 540 Hz} \\ \\ Equivalent Input Noise = 0.24 \ \mu\text{V}_{rms} \end{array}$$

Note: Indicates  $\pm 1\%$  metal film resistors recommended for temperature stability.

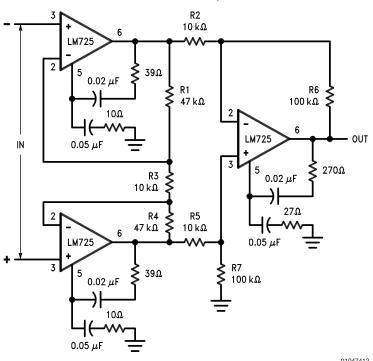
## Typical Applications (Continued)

#### ±100V Common Mode Range Differential Amplifier



## Typical Applications (Continued)

#### Instrumentation Amplifier with High **Common Mode Rejection**



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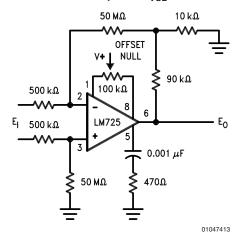
$$\frac{R1}{R6} = \frac{R3}{R4} \text{ for best CMRR}$$

$$R3 = R4$$

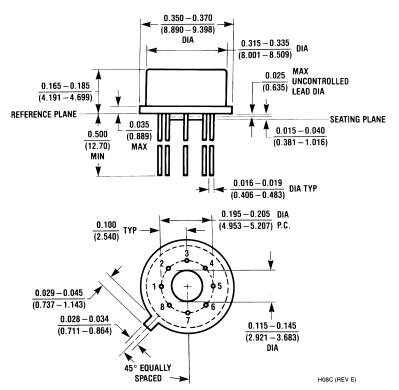
$$R1 = R6 = 10 R3$$

$$Gain = \frac{R6}{R7}$$

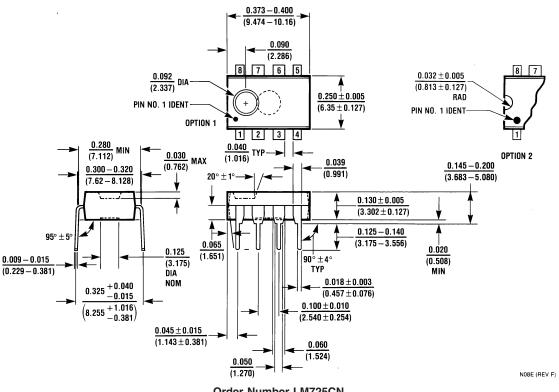
### Precision Amplifier $A_{VCL} = 1000$



# **Physical Dimensions** inches (millimeters) unless otherwise noted



Order Number LM725H/883, LM725CH or LM725AH/883 **NS Package Number H08C** 



**Order Number LM725CN NS Package Number N08E** 

#### **Notes**

National does not assume any responsibility for use of any circuitry described, no circuit patent licenses are implied and National reserves the right at any time without notice to change said circuitry and specifications.

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Email: new.feedback@nsc.com Tel: 1-800-272-9959

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Email: europe.support@nsc.com
Deutsch Tel: +49 (0) 69 9508 6208
English Tel: +44 (0) 870 24 0 2171
Français Tel: +33 (0) 1 41 91 8790

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